

40V, 5MHz, Rail-to-Rail Input/Output Single Supply Operational Amplifiers

Features

- Operates on
Single-Supply: 4.5V ~ 40V, or
Dual-Supply: $\pm 2.25\text{V} \sim \pm 20\text{V}$
- Low Quiescent Current: 700 μA
- Gain Bandwidth Product: 5MHz
- Rail-to-Rail Input and Output (RRIO)
- 3.0V/ μs Slew Rate
- Low offset drift: $\pm 0.3\mu\text{V}/^\circ\text{C}$ (Typical)
- Unity Gain Stable
- No Phase Reversal
- RFI Filtered Inputs
- Low-input Bias current: 10pA
- Extended Temperature Ranges
From -40°C to $+125^\circ\text{C}$
- Small Packaging
COS171 available in SOT23-5/SOP-8
COS2171 available in SOP-8/MSOP-8
COS4171 available in SOP14/TSSOP14

Applications

- Automotive Electronics
- Sensor Conditioning
- AGC Control
- Tracking Amplifier in Power Modules
- Battery-Powered Instruments
- Test Equipment

General Description

The COS171 (single), COS2171 (dual) and COS4171 (quad) are single supply, rail-to-rail input and output amplifiers operated on 4.5V($\pm 2.25\text{V}$) to 40V($\pm 20\text{V}$) supplies. Despite their low quiescent current, the COSx171 family provides excellent overall performance and versatility.

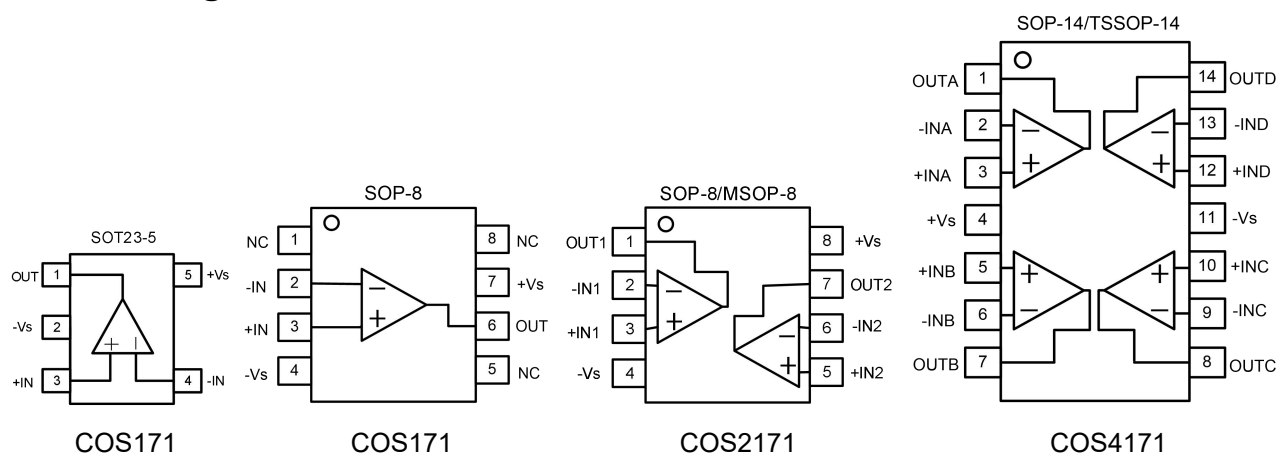
COSx171 family is unity gain stable and has a gain bandwidth product of 5MHz (typical). They provide high CMRR and PSRR performance and can operate from a single supply or dual supply. Input signals can go beyond the supply rails and do not cause phase reversal. The output voltage swing extends to within 150mV of each rail, providing the maximum output dynamic range with excellent overdrive recovery. These features make the COSx171 family well suited for single-supply, battery-powered applications. They can be used as plus-in replacements for many commercially available op-amps to reduce power and improve input/output range and performance.

Rev1.2

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1. Pin Configuration and Functions



Pin Functions

Name	Description	Note
+Vs	Positive power supply	A bypass capacitor of 0.1μF as close to the part as possible should be placed between power supply pins or between supply pins and ground.
-Vs	Negative power supply or ground	If it is not connected to ground, bypass it with a capacitor of 0.1μF as close to the part as possible.
-IN	Negative input	Inverting input of the amplifier. Voltage range of this pin can go from -Vs -0.3V to +Vs + 0.3V.
+IN	Positive input	Non-inverting input of the amplifier. This pin has the same voltage range as -IN.
OUT	Output	The output voltage range extends to within millivolts of each supply rail.
NC	No connection	No internal connection, can be left floating

2. Package and Ordering Information

Model	Channel	Order Number	Package	Package Option	Marking Information
COS171	1	COS171TR	SOT23-5	Tape and Reel, 3000	COS171
		COS171TRB			COS171
		COS171SR	SOP-8	Tape and Reel, 4000	COS171SR
		COS171SRB			COS171SR
COS2171	2	COS2171SRA	SOP-8	Tape and Reel, 4000	COS2171SR
		COS2171SRB			COS2171SR
		COS2171MR	MSOP-8	Tape and Reel, 3000	COS2171MR
		COS2171MRB			COS2171MR

COS4171	4	COS4171SR	SOP-14	Tape and Reel, 2500	COS4171SR
		COS4171TR	TSSOP-14	Tape and Reel, 3000	COS4171TR

3. Product Specification

3.1 Absolute Maximum Ratings ⁽¹⁾

Parameter	Rating	Units
Power Supply: +Vs to -Vs	42	V
Input Voltage	-Vs -0.5V to +Vs + 0.5V	V
Input Current ⁽²⁾	±10	mA
Storage Temperature Range	-65 to 150	°C
Junction Temperature	150	°C
Operating Temperature Range	-40 to 125	°C
ESD Susceptibility, HBM	2000	V

(1) Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

(2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.

3.2 Thermal Data

Parameter	Rating	Unit
Package Thermal Resistance, R _{θJA} (Junction-to-ambient)	190 (SOT23-5) 206 (MSOP8) 155 (SOP8) 105 (TSSOP14) 82 (SOP14)	°C/W

3.3 Recommended Operating Conditions

Parameter	Rating	Unit
DC Supply Voltage	4.5 (±2.25) ~ 40 (±20)	V
Input common-mode voltage range	-Vs ~ +Vs	V
Operating ambient temperature	-40 to +125	°C

3.4 Electrical Characteristics

($V_S=4.5$ to $40V$, $V_{CM}=V_{OUT}=V_S/2$, $T_A=+25^{\circ}C$, $R_L=10k\Omega$ to $V_S/2$, unless otherwise noted)

Parameter	Symbol	Conditions	Min	Typ	Max	Unit
Input Characteristics						
Input Offset Voltage	V_{OS}	COS171, COS2171A		± 0.3	± 2.0	mV
		COS171B, COS2171B		± 0.75	± 5.0	
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta T$	-40 to $125^{\circ}C$		± 0.3	± 3.0	$\mu V/^{\circ}C$
Input Bias Current	I_B			± 10		pA
Input Offset Current	I_{OS}			± 4		pA
Common-Mode Voltage Range	V_{CM}	$V_S = 40V$	V_{-S} -0.1		V_{+S} -2	V
Common-Mode Rejection Ratio	CMRR	$V_{-S} - 0.1 < V_{CM} < V_{+S} - 2$	100	120		dB
Open-Loop Voltage Gain	AOL	$V_O=0.2V$ to $4.8V$	110	130		dB
Output Characteristics						
Output Voltage Swing from Rail		$R_L=10k\Omega$		150		mV
Short-Circuit Current	I_{SR}	Sourcing		+40		mA
	I_{SK}	Sinking		-48		mA
Power Supply						
Operating Voltage Range			4.5		40	V
Power Supply Rejection Ratio	PSRR	$V_S = +4.5V$ to $+40V$	80	100		dB
Quiescent Current / Amplifier	I_Q			700		μA
Dynamic Performance						
Gain Bandwidth Product	GBWP	$G=+1$		5.0		MHz
Slew Rate	SR	$G=+1$		3.0		V/ μs
Noise Performance						
Voltage Noise Density	e_n	$f=1kHz$		18		nV/ \sqrt{Hz}

4.0 Application Notes

Driving Capacitive Loads

Driving large capacitive loads can cause stability problems for voltage feedback op amps. As the load capacitance increases, the feedback loop's phase margin decreases, and the closed loop bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in the step response. A unity gain buffer ($G = +1$) is the most sensitive to capacitive loads, but all gains show the same general behavior.

When driving large capacitive loads with these op amps (e.g., > 100 pF when $G = +1$), a small series resistor at the output (R_{ISO} in Figure 1) improves the feedback loop's phase margin (stability) by making the output load resistive at higher frequencies. It does not, however, improve the bandwidth.

To select R_{ISO} , check the frequency response peaking (or step response overshoot) on the bench. If the response is reasonable, you do not need R_{ISO} . Otherwise, start R_{ISO} at $1\text{ k}\Omega$ and modify its value until the response is reasonable.

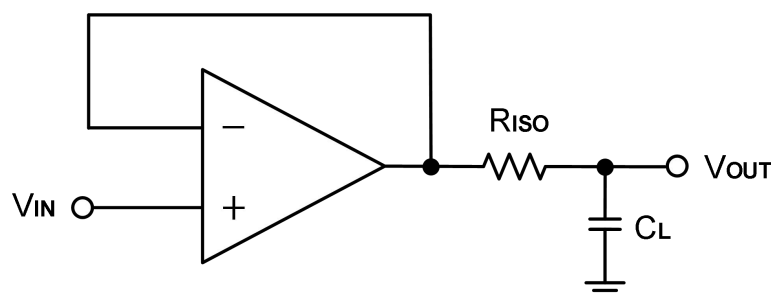


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2. It provides DC accuracy as well as AC stability. R_F provides the DC accuracy by connecting the inverting signal with the output, C_F and R_{ISO} serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

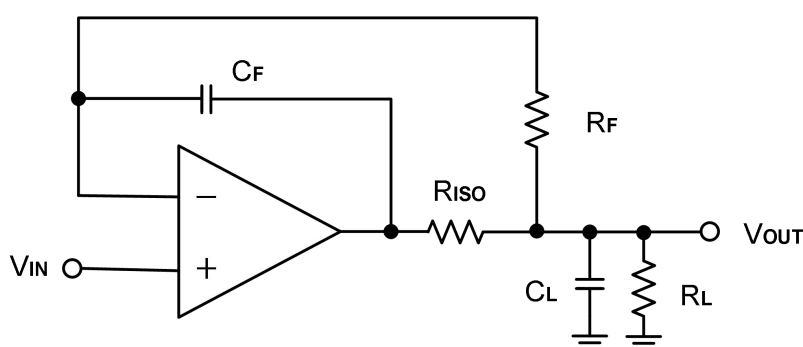


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

For inverting configuration, there are two others ways to increase the phase margin: (a) by increasing the amplifier's gain or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node, as shown in Figure 3.

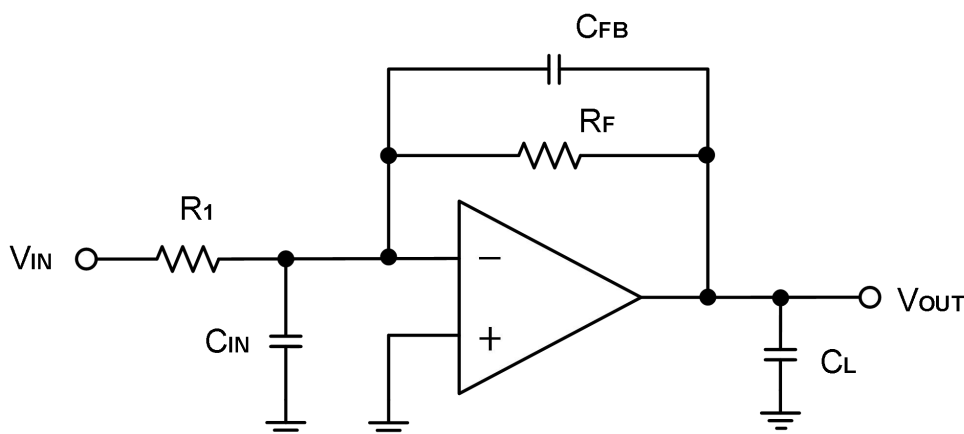


Figure 3. Adding a Feedback Capacitor in the Inverting Configuration

Power-Supply Bypassing and Layout

The COSx171 family operates from a single +4.5V to +40V supply or dual $\pm 2.25\text{V}$ to $\pm 20\text{V}$ supplies. For single-supply operation, bypass the power supply +Vs with a $0.1\mu\text{F}$ ceramic capacitor which should be placed close to the +Vs pin. For dual-supply operation, both the +Vs and the -Vs supplies should be bypassed to ground with separate $0.1\mu\text{F}$ ceramic capacitors. $2.2\mu\text{F}$ tantalum capacitor can be added for better performance.

The length of the current path is directly proportional to the magnitude of parasitic inductances and thus the high frequency impedance of the path. High speed currents in an inductive ground return create an unwanted voltage noise. Broad ground plane areas will reduce the parasitic inductance. Thus a ground plane layer is important for high speed circuit design.

Typical Application Circuits

Differential Amplifier

The circuit shown in Figure 4 performs the differential function. If the resistors ratios are equal ($R_4 / R_3 = R_2 / R_1$), then $V_{OUT} = (V_{IP} - V_{IN}) \times R_2 / R_1 + V_{REF}$.

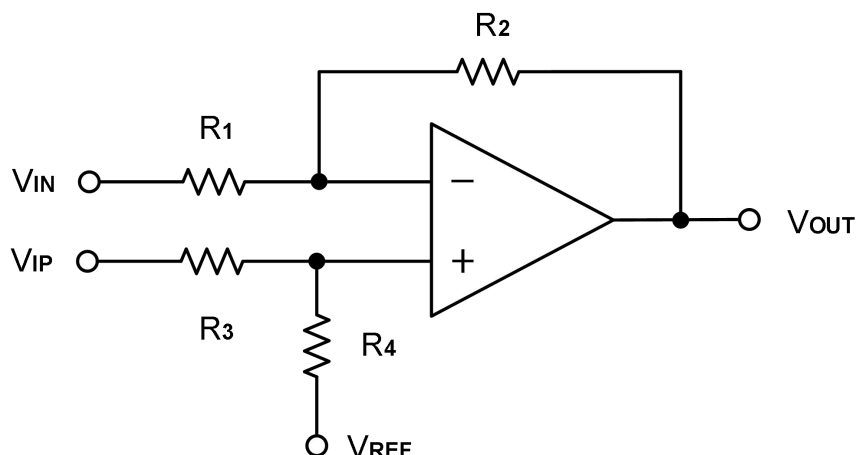


Figure 4. Differential Amplifier

Low Pass Active Filter

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to establish this limited bandwidth is to place an RC filter at the noninverting terminal of the amplifier. If even more attenuation is needed, a multiple pole filter is required. The Sallen-Key filter can be used for this task, as Figure 5. For best results, the amplifier should have a bandwidth that is 8 to 10 times the filter frequency bandwidth. Failure to follow this guideline can result in reduction of phase margin. The large values of feedback resistors can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistors value as low as possible and consistent with output loading consideration.

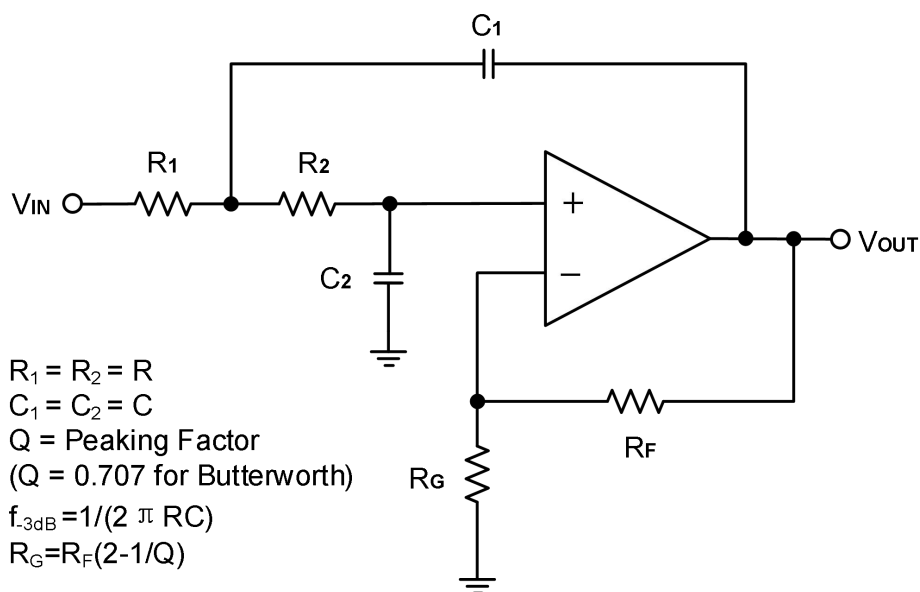
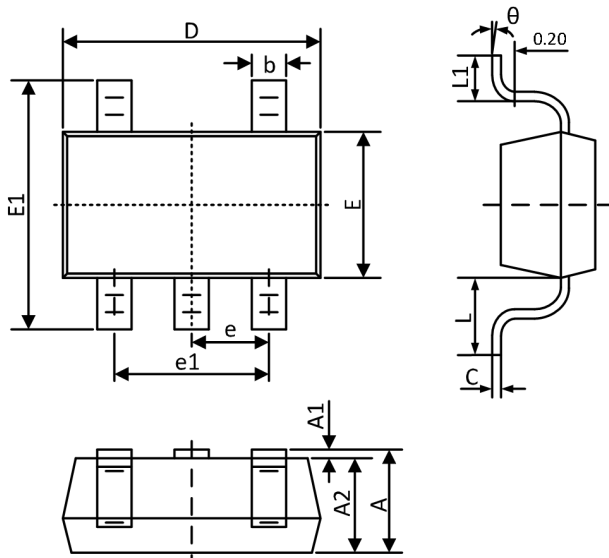


Figure 5. Two-Pole Low-Pass Sallen-Key Active Filter

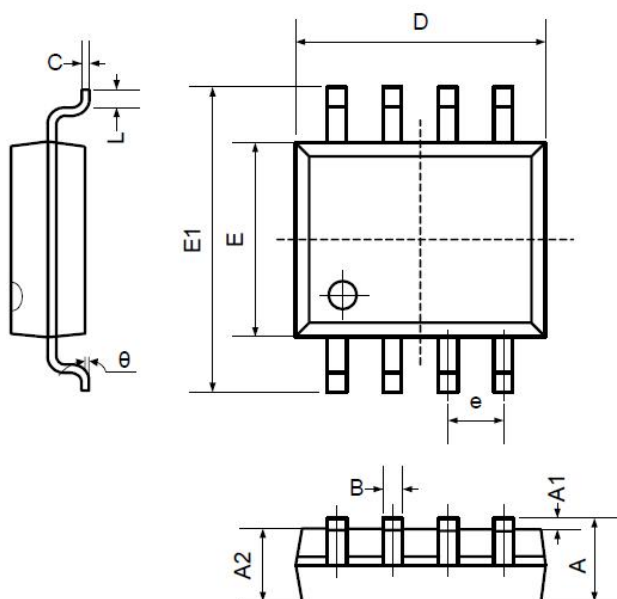
5. Package Information

5.1 SOT23-5 (Package Outline Dimensions)



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	MIN	MAX	MIN	MAX
A	1.050	1.250	0.041	0.049
A1	0.000	0.100	0.000	0.004
A2	1.050	1.150	0.041	0.045
b	0.300	0.400	0.012	0.016
c	0.100	0.200	0.004	0.008
D	2.820	3.020	0.111	0.119
E	1.500	1.700	0.059	0.067
E1	2.650	2.950	0.104	0.116
e	0.950TYP		0.037TYP	
e1	1.800	2.000	0.071	0.079
L	0.700REF		0.028REF	
L1	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°

5.2 SOP8 (Package Outline Dimensions)



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
B	0.330	0.510	0.013	0.020
C	0.190	0.250	0.007	0.010
D	4.780	5.000	0.188	0.197
E	3.800	4.000	0.150	0.157
E1	5.800	6.300	0.228	0.248
e	1.270TYP		0.050TYP	
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°

Technical drawing of a rectangular component with four pins, showing top, side, and detail views with dimensions:

- Top View:** Shows a rectangular body with a central hole. Dimensions include overall width $E1$, body width E , pin spacing b , and pin width e .
- Side View:** Shows the component's profile with dimensions $A1$ (total height), A (body height), and $A2$ (pin height).
- Detail View:** Shows a close-up of the pin and body interface with dimensions C (fillet radius) and θ (fillet angle).

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.800	1.200	0.031	0.047
A1	0.000	0.200	0.000	0.008
A2	0.760	0.970	0.030	0.038
b	0.30 TYP		0.012 TYP	
c	0.15 TYP		0.006 TYP	
D	2.900	3.100	0.114	0.122
e	0.65 TYP		0.026 TYP	
E	2.900	3.100	0.114	0.122
E1	4.700	5.100	0.185	0.201
L	0.410	0.650	0.016	0.026
θ	0°	6°	0°	6°

Technical drawing of a 10-pin D-subminiature connector, showing top, side, and cross-sectional views with dimensions and tolerances.

Top View:

- Overall width: D
- Overall height: E
- Pin pitch (center-to-center): e
- Pin width: b
- Pin diameter: $\phi 0.25 (M)$
- Index hole diameter: $\phi 0.8 \pm 0.1$
- Deposition thickness: $DEP0.2 \pm 0.1$
- Internal hole diameter: $\phi 2.0 \pm 0.1$ BTM E-MARK
- Internal hole deposition thickness: $DEP0.1 \pm 0.05$

Side View:

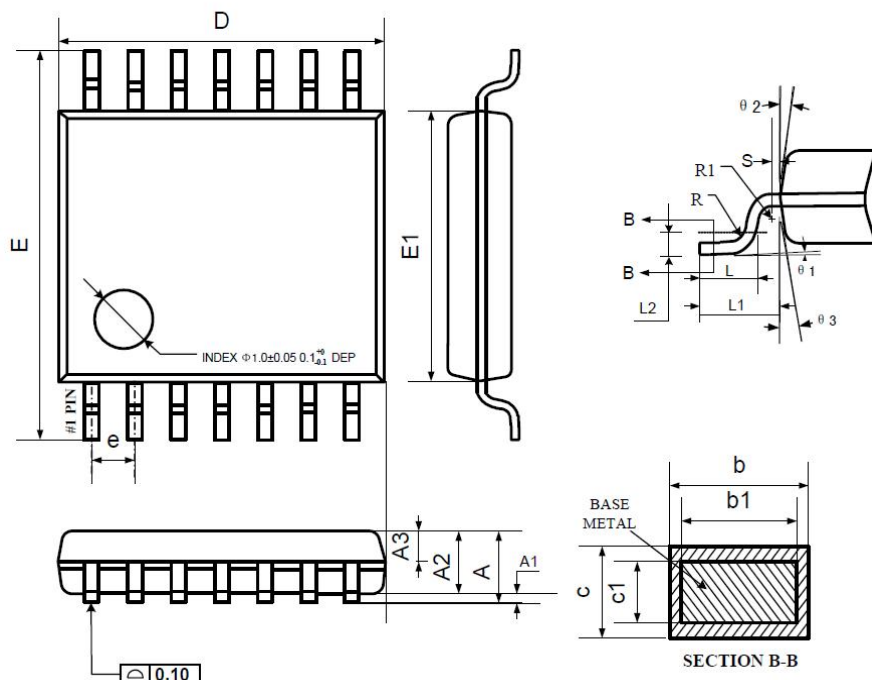
- Overall height: A
- Pin height: $A2$
- Pin diameter: $\phi 3$
- Pin width: $\phi 4$
- Pin thickness: $A3$
- Pin diameter tolerance: $\phi 0.10$

Cross-sectional View (SECTION B-B):

- Overall width: b
- Pin width: $b1$
- Overall height: c
- Pin height: $c1$
- Labels: BASE METAL, WITH PLATING

Symbol	Dimensions In Millimeters		
	MIN	NOM	MAX
A	1.35	1.60	1.75
A1	0.10	0.15	0.25
A2	1.25	1.45	1.65
A3	0.55	0.65	0.75
b	0.36		0.49
b1	0.35	0.40	0.45
c	0.16		0.25
c1	0.15	0.20	0.25
D	8.53	8.63	8.73
E	5.80	6.00	6.20
E1	3.80	3.90	4.00
e		1.27 BSC	
L	0.45	0.60	0.80
L1		1.04 REF	
L2		0.25 BSC	
R	0.07		
R1	0.07		
h	0.30	0.40	0.50
θ	0°		8°
θ1	6°	8°	10°
θ2	6°	8°	10°
θ3	5°	7°	9°
θ4	5°	7°	9°

5.5 TSSOP14 (Package Outline Dimensions)



Symbol	Dimensions In Millimeters		
	MIN	NOM	MAX
A	—	—	1.20
A1	0.05	—	0.15
A2	0.90	1.00	1.05
A3	0.34	0.44	0.54
b	0.20	—	0.28
b1	0.20	0.22	0.24
c	0.10	—	0.19
c1	0.10	0.13	0.15
D	4.86	4.96	5.06
E	6.20	6.40	6.60
E1	4.30	4.40	4.50
e	0.65 BSC		
L	0.45	0.60	0.75
L1	1.00 REF		
L2	0.25 BSC		
R	0.09	—	—
R1	0.09	—	—
S	0.20	—	—
$\theta 1$	0°	—	8°
$\theta 2$	10°	12°	14°
$\theta 3$	10°	12°	14°

6. Related Parts

Part Number	Description
COS6042	24kHz, 0.5 μ A, Nano-Power Op Amps, 1.4V to 5.5V Supply
COS8042	160MHz, 5.5mA, High Speed Op Amps, 3V to 12V Supply
COS2172	10MHz, 1.2mA, RRIO Op Amps, 4.5 to 40V Supply
COS2333	350kHz, 18 μ A, Precision Op Amps, 1.8 to 5.5V Supply, Zero Drift, Vos<10 μ V
COS8552	1.5MHz, 55 μ A, Precision Op Amps, 1.8 to 5.5V Supply, Zero Drift, Vos<10 μ V
COS2388	9MHz, 570 μ A, Precision Op Amps, 1.8 to 5.5V Supply, Zero Drift, Vos<10 μ V
COS2227	10MHz, 1.3mA, Precision Op Amps, 4.5 to 36V Supply, Vos<50 μ V
COS2182	5MHz, 580 μ A, RRIO Precision Op Amps, 4.5 to 40V Supply, Vos<50 μ V
COS620	1.5MHz, 1.3mA, Instrumentation Amps, 4.5 to 36V Supply, Vos<50 μ V
COSINA333	150kHz, 65 μ A, Instrumentation Amps, 1.8 to 5.5V Supply, Vos<25 μ V