

# 60ns, 35μA, RRI, Push-Pull Output Comparators

#### **Features**

Low Propagation Delay: 60 ns

■ Low Quiescent Current: 35µA per Channel

■ Low Input Offset Voltage: 1 mV

■ Supply Operation From: 1.8V ~ 5.5V

■ Rail-to-Rail Inputs

Push-Pull Outputs

■ Internal Hysteresis for Noise Immunity

Input Bias Current: 1.0 pA Typical

No Phase Inversion for Overdrive Inputs

Extended Temperature Ranges
 From -40°C to +125°C

Small Packaging: MSOP8 and SOP-8

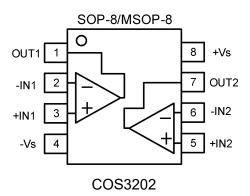
## **Applications**

- Inspection Equipment
- Threshold Detectors/Discriminators
- Peak and Zero-crossing detectors
- Logic Level Shifting or Translation
- Portable Equipment
- Sensor Conditioning
- High-Speed Sampling Systems
- Battery Powered Electronics
- IR Receivers
- Window Comparators

## **General Description**

The COS3202 is a dual-channel low-power comparator which features high speed time with rail-to-rail inputs. It response 35µA per Channel while only achieving a 60ns propagation delay. The operating voltage ranges from 1.8 V to 5.5 V. A 1.2mV internal hysteresis is built in to improved noise immunity.

The COS3202 has push-pull output stages capable of sinking and sourcing milliamps of current when controlling an LED or driving a capacitive load. The COS3202 is available in small 8-pin MSOP-8 and SOP-8 packages.

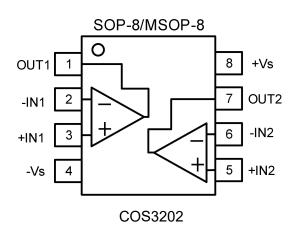


Rev1.0

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# 1. Pin Configuration and Functions



# **Pin Functions**

PIN No.	Name	Туре	Description	
1	OUT1	0	Output, Comparator1	
2	-IN1	I	Negative input, Comparator1	
3	+IN1	I	Positive input, Comparator1	
4	-Vs	Р	Negative power supply or ground	
5	+IN2	I	Positive input, Comparator2	
6	-IN2	I	Negative input, Comparator2	
7	OUT2	0	Output, Comparator2	
8	+Vs	Р	Positive power supply	

# 2. Package and Ordering Information

Model	Channel	Order Number	Package	Package Option	Marking Information
COS3202	3202 2	COS3202SR	SOP-8	Tape and Reel, 4000	COS3202
CO33202		COS3202MR	MSOP-8	Tape and Reel, 3000	COS3202



## 3. Product Specification

## 3.1 Absolute Maximum Ratings (1)

Parameter	Rating	Units
Power Supply: +Vs to -Vs	6	V
Input Voltage	-Vs -0.5V to +Vs + 0.5V	V
Input Current (2)	±10	mA
Output Short Circuit Current	±70	mA
Storage Temperature Range	-65 to 150	°C
Junction Temperature	150	°C
Operating Temperature Range	-40 to 125	°C
ESD Susceptibility, HBM	3000	V

<sup>(1)</sup> Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

#### 3.2 Thermal Data

Parameter	Rating	Unit
Package Thermal Resistance, R <sub>θJA</sub> (Juntion-to-ambient)	206 (MSOP8) 155 (SOP8)	°C/W

### 3.3 Recommended Operating Conditions

Parameter	Rating	Unit
DC Supply Voltage	1.8 (±0.9) ~ 5.5 (±2.75)	V
Input Common-mode Voltage Range	-Vs ~ +Vs	V
Operating Ambient Temperature	-40 ~ +125	°C

<sup>(2)</sup> Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.



## 3.4 Electrical Characteristics

( $V_S = 5V$ ,  $C_L = 15pF$ ,  $T_A = +25$ °C, unless otherwise noted)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Input Characteristics						
Input Offset Voltage	Vos			1.0	5.0	mV
Input Offset Voltage Drift	ΔV <sub>OS</sub> /ΔΤ	-40 to 125°C		2	10	μV/°C
Input Hysteresis	V <sub>HYS</sub>			1.2		mV
Input Bias Current	I <sub>B</sub>			1.0	10	pА
Input Offset Current	los			1.0	10	рА
Common-Mode Voltage Range	V <sub>CM</sub>	-40 to 125°C	-0.2		Vs+0.2	V
Common-Mode Rejection Ratio	CMRR	-0.2 V < V <sub>CM</sub> < 5.2 V	60	75		dB
Output Characteristics						
Voltage Output Swing from Lower Rail	V <sub>OL</sub>	I <sub>SINK</sub> =4mA		154		mV
Voltage Output Swing from Upper Rail	Vон	I <sub>SOURCE</sub> =4mA		150		mV
Short-Circuit Current	I <sub>SR</sub>	Sourcing		58		mA
Short-Circuit Current	I <sub>SK</sub>	Sinking		59		mA
Power Supply						
Operating Voltage Range	Vs		1.8		5.5	V
Power Supply Rejection Ratio	PSRR	V <sub>S</sub> = +2.5V to +5.5V	70	90		dB
Quiescent Current (nor Channel)		V <sub>S</sub> = +5.0V		35	45	μΑ
Quiescent Current (per Channel)	l IQ	-40 to 125°C			50	μΑ
Switching Characteristics						
Propagation Delay Time, High to Low	t <sub>PHL</sub>	V <sub>OD</sub> = 100mV		60	75	ns
Propagation Delay Time, Low to High	t <sub>PLH</sub>	V <sub>OD</sub> = 100mV		60	75	ns
Rise Time	t <sub>R</sub>	10% to 90%		5		ns
Fall Time	t <sub>F</sub>	90% to 10%		5		ns



## 4.0 Application Notes

#### **Inverting Comparator with Hysteresis**

When higher levels of hysteresis are required, positive feedback can be externally added. The inverting comparator with hysteresis requires a three-resistor network that is referenced to the comparator supply voltage (Vcc), as shown in Figure 1. When  $V_{IN}$  at the inverting input is less than  $V_{T}$ , the output voltage is high. The three network resistors can be represented as R1//R3 in series with R2. Equation 1 defines the high to low trip voltage ( $V_{T1}$ ).

$$V_{T1} = \frac{R2 \cdot Vcc}{(R1//R3) + R2} \tag{1}$$

When  $V_{IN}$  is greater than  $V_A$ , the output voltage is low, very close to ground. In this case, the three network resistors can be presented as R2//R3 in series with R1. Equation 2 define the low to high trip voltage ( $V_{T2}$ ).

$$V_{T2} = \frac{(R2//R3) \cdot Vcc}{(R2//R3) + R1} \tag{2}$$

The total hysteresis provided by the network is

$$\Delta V_{\rm T} = V_{T1} - V_{T2} \tag{3}$$

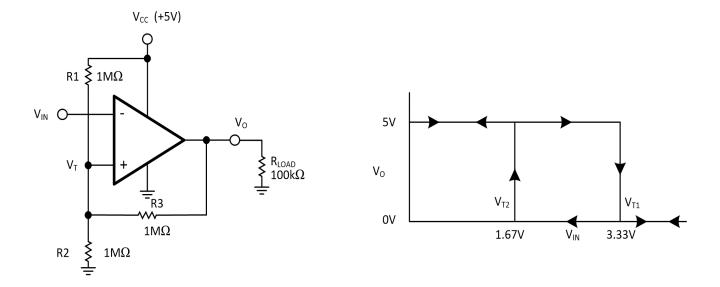


Figure 1. Inverting Configuration with Hysteresis



#### Non-inverting Comparator with Hysteresis

A non-inverting comparator with hysteresis requires a two-resistor network, as shown in Figure 2, and a voltage reference ( $V_{REF}$ ) at the inverting input. When  $V_{IN}$  is low, the output is also low. For the output to switch from low to high,  $V_{IN}$  must rise to  $V_{IN1}$ . Equation 4 defines the low to high trip voltage ( $V_{IN1}$ ):

$$V_{IN1} = \frac{(R1+R2) \cdot V_{REF}}{R2} \tag{4}$$

When  $V_{IN}$  is high, the output is also high. For the comparator to switch back to a low state,  $V_{IN}$  must drop to  $V_{IN2}$ ,

$$V_{IN2} = \frac{(R1+R2) \cdot V_{REF} - R1 \cdot V_{CC}}{R2} \tag{5}$$

The hysteresis of this circuit is the difference between  $V_{\text{IN1}}$  and  $V_{\text{IN2}}$ , as shown in following,

$$\Delta V_{\rm IN} = V_{CC} \times \frac{R1}{R2} \tag{6}$$

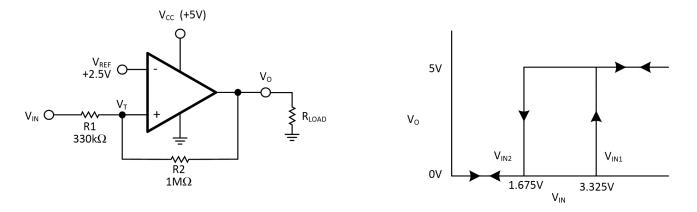


Figure 2. Non-inverting Configuration with Hysteresis

#### Square-Wave Oscillator

The COS3202 can be used to build a low cost square-wave oscillator as shown in Figure 3. The square-wave period is determined by the RC time constant of the capacitor (C1) and resistor (R4). The maximum frequency is limited by propagation delay of the device and the capacitance load at the output.



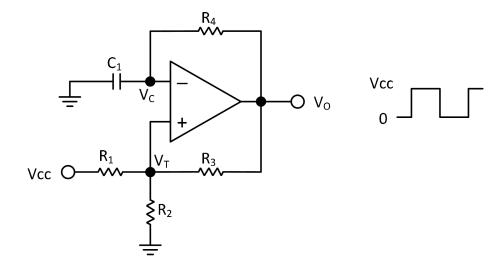


Figure 3. Square-Wave Oscillator

#### **IR Receiver**

COS3202 can be used to build a IR receiver analog front end as shown in Figure 4. R1 converts the IR light energy induced current into voltage and applies to the inverting input of the comparator. The RC network of R2 and C1 establishes a reference voltage Vref which tracks the mean amplitude of the IR signal. The RC constant of R2 and C1 is chosen for Vref to track the received IR current fluctuation but not the actual data bit stream. The non-inverting input is connected to Vref and the output over the R3 and R4 resistor network which provides hysteresis for improved guard against spurious toggles.

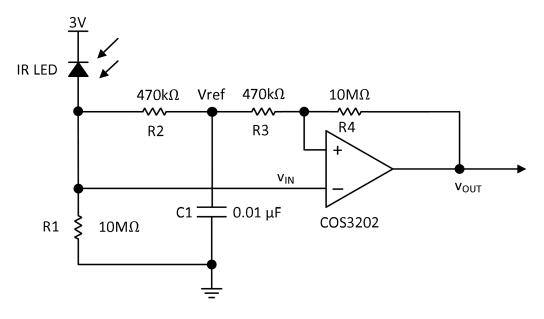


Figure 4. IR Receiver



#### **Power-Supply Bypassing and Layout**

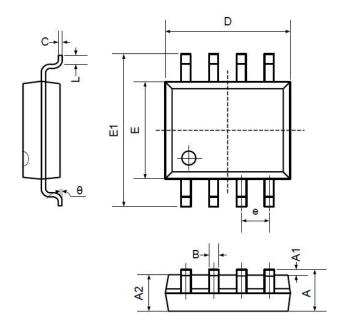
For single-supply operation, bypass the power supply  $V_S$  with a  $0.1\mu F$  ceramic capacitor which should be placed close to the  $V_S$  pin. For dual-supply operation, both the positive and negative supplies should be bypassed to ground with separate  $0.1\mu F$  ceramic capacitors.  $2.2\mu F$  tantalum capacitor can be added for better performance.

The length of the current path is directly proportional to the magnitude of parasitic inductances and thus the high frequency impedance of the path. High speed currents in an inductive ground return create an unwanted voltage noise. Broad ground plane areas will reduce the parasitic inductance. Thus a ground plane layer is important for high speed circuit design.



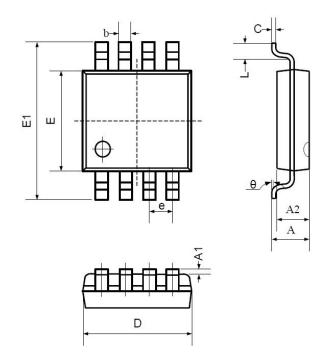
# 5. Package Information

# 5.1 SOP8 (Package Outline Dimensions)



Symbol		nsions meters	Dimensions In Inches		
	Min	Max	Min	Max	
Α	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
В	0.330	0.510	0.013	0.020	
С	0.190	0.250	0.007	0.010	
D	4.780	5.000	0.188	0.197	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.300	0.228	0.248	
е	1.270TYP		0.050TYP		
L	0.400	1.270	0.016	0.050	
θ	0°	8°	0°	8°	

## **5.2 MSOP8 (Package Outline Dimensions)**



Symbol	Dimensions In Millimeters		Dimensions In Inches		
	Min	Max	Min	Max	
Α	0.800	1.200	0.031	0.047	
A1	0.000	0.200	0.000	0.008	
A2	0.760	0.970	0.030	0.038	
b	0.30 TYP		0.012 TYP		
С	0.15	0.15 TYP		0.006 TYP	
D	2.900	3.100	0.114	0.122	
е	0.65 TYP		0.026 TYP		
E	2.900	3.100	0.114	0.122	
E1	4.700	5.100	0.185	0.201	
L.	0.410	0.650	0.016	0.026	
θ	0°	6°	0°	6°	